# Time-Domain Analysis of Transmission Line Circuits

(Part 1)

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# Outline

- Quarter-wave transformer steady state response
- Quarter-wave transformer transient response
- Reflection coefficient revised
- Concept of "transient impedance"
- Applying DC to transmission lines
- Lattice (or bouncing or reflection) diagrams
- Building transient signals from bouncing diagrams

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 $Z_{in} = \frac{Z_1^2}{R_L}$ To make  $\Gamma = 0$ ,  $Z_1 = \sqrt{R_L Z_o}$ Z must be the

 $Z_1$  must be the geometric mean of  $Z_o$  and  $R_L$ 

 $\Gamma$  is the steadystate reflection coefficient

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Quarter-Wave Transformer – Transient Response  $\Gamma_{1} = \frac{Z_{1} - Z_{o}}{Z_{1} + Z_{o}}$   $\Gamma_{2} = \frac{Z_{o} - Z_{1}}{Z_{o} + Z_{1}} = -\Gamma_{1}$   $\Gamma_{3} = \frac{R_{L} - Z_{1}}{R_{L} + Z_{1}}$   $T_{1} = 1 + \Gamma_{1} = \frac{2Z_{1}}{Z_{1} + Z_{o}}$   $T_{2} = 1 + \Gamma_{2} = \frac{2Z_{o}}{Z_{o} + Z_{1}}$ 

(D. M. Pozar, Microwave Engineering, Wiley, 2005) 4

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Dr. J. E. Rayas Sánchez http://iteso.mx/~erayas erayas@iteso.mx Quarter-Wave Transformer – Transient Response



Initially, the incoming wave has only the incident component:  $V(z) = V_o^+ e^{-j\beta z}$ 

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Quarter-Wave Transformer - Transient Response Ζ. My Voteria  $V_0^+$  $\mathcal{N}$  $\lambda/4$  $T_1 V_0^+$ T1V0+CU V0+e-jn/2  $\leftarrow \int \Gamma_3 T_1 V_0^+ e^{-j\pi/2}$  $-\Gamma_3T_1T_2V_0^+$  ${}_{3}T_{1}V_{0}$  $\Gamma_2\Gamma_3T_1V_0^+e^{-j\pi/2}$ -iπ2  $-\Gamma_2\Gamma_3^2T_1V_0$  $\Gamma_2\Gamma_3^2T_1T_2V_0^+$ Dr. J.E. Rayas Sánchez

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Quarter-Wave Transformer - Transient Response



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Quarter-Wave Transformer - Transient Response

$$\begin{split} &\Gamma = \Gamma_{1} - T_{1}T_{2}\Gamma_{3}\sum_{n=0}^{\infty}(-\Gamma_{2}\Gamma_{3})^{n} \\ &\text{Since }\sum_{n=0}^{\infty}x^{n} = \frac{1}{1-x}, \text{ for } |x| < 1 \\ &\Gamma = \Gamma_{1} - \frac{T_{1}T_{2}\Gamma_{3}}{1+\Gamma_{2}\Gamma_{3}} = \frac{\Gamma_{1} + \Gamma_{1}\Gamma_{2}\Gamma_{3} - T_{1}T_{2}\Gamma_{3}}{1+\Gamma_{2}\Gamma_{3}} \\ &\text{using} \\ &\Gamma_{1} = \frac{Z_{1} - Z_{o}}{Z_{1} + Z_{o}} \quad \Gamma_{2} = -\Gamma_{1} \quad \Gamma_{3} = \frac{R_{L} - Z_{1}}{R_{L} + Z_{1}} \quad T_{1} = \frac{2Z_{1}}{Z_{1} + Z_{o}} \quad T_{2} = \frac{2Z_{o}}{Z_{o} + Z_{1}} \\ &\Gamma_{1} + \Gamma_{1}\Gamma_{2}\Gamma_{3} - T_{1}T_{2}\Gamma_{3} = \frac{2(Z_{1}^{2} - Z_{o}R_{L})}{(Z_{1} + Z_{o})(R_{L} + Z_{1})} = 0 \text{ if } Z_{1} = \sqrt{Z_{o}R_{L}} \end{split}$$

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Quarter-Wave Transformer – Transient Response  $\begin{array}{c} & & \\ \hline & & \\ \hline & & \\ Z_0 \\ & & \\ Z_0 \\ & & \\ Z_1 \\ & \\ Z_1 \\ & \\ R_L \\ & \\ R_L \\ & \\ F = \Gamma_1 - T_1 T_2 \Gamma_3 \sum_{n=0}^{\infty} (-\Gamma_2 \Gamma_3)^n = 0 \quad \text{if } Z_1 = \sqrt{Z_o R_L}
\end{array}$ 

If  $Z_1$  is the geometric mean of  $Z_o$  and  $R_L$ , the sum of the infinite number of partial reflections is zero

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Applying DC to Transmission Lines (cont)



$$\Gamma_{s} = \frac{R_{s} - Z_{o}}{R_{s} + Z_{o}} \qquad \Gamma_{L} = \frac{R_{L} - Z_{o}}{R_{L} + Z_{o}} \qquad v_{p} = \frac{c}{\sqrt{\varepsilon_{e}}}$$
$$V_{o}^{+} = \frac{V_{s}Z_{o}}{R_{s} + Z_{o}} \qquad I_{o}^{+} = \frac{V_{o}^{+}}{Z_{o}}$$

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# Lattice (or Bouncing or Reflection) Diagrams

Voltage Diagrams



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Bouncing Diagrams for Currents



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(M. Leddige, 2003)

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### Example – Underdriven Transmission Line

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### Example – Underdriven Transmission Line (cont)



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#### Example – Overdriven Transmission Line

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$$v(x=0,t) = ?; v(x=l,t) = ?; v(x=\frac{3}{4}l,t) = ?$$
  
 $i(x=0,t) = ?; i(x=l,t) = ?; i(x=\frac{3}{4}l,t) = ?$ 

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Example Simulated with APLAC





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