

## **Impedance Matching Circuits (Part 2)**

Dr. José Ernesto Rayas-Sánchez

April 23, 2020

# **Impedance Matching Circuits**

(Part 2)

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## **Outline**

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- Capacitive Loads
- Resistive compensation of capacitive loads
- Inductive compensation of capacitive loads
- Matching capacitive loads with an inductor
- Matching capacitive loads with L-sections
- Other practical techniques

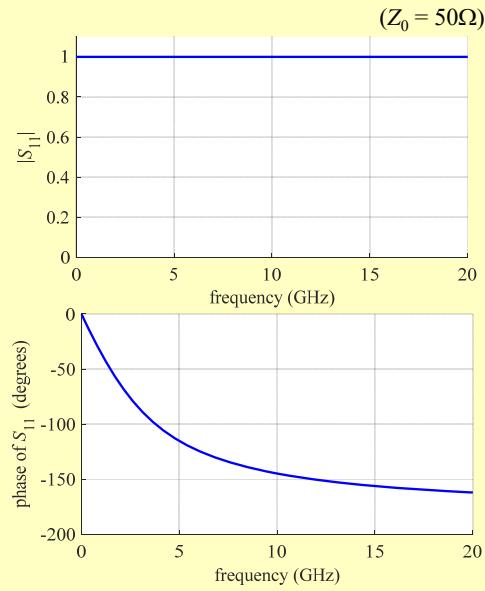
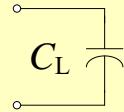
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### Capacitive Load – Example

$$C_L = 1\text{pF}$$

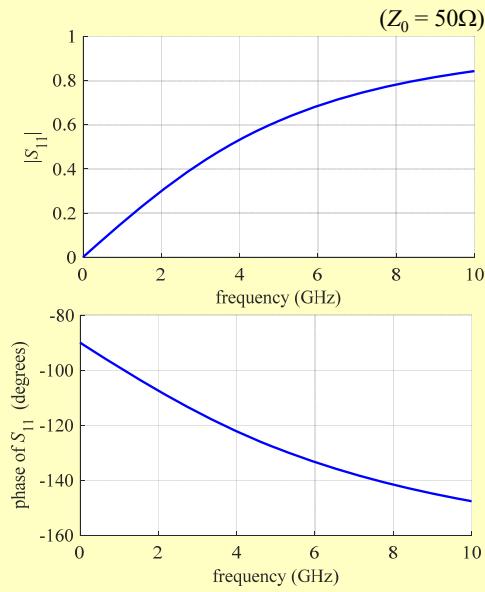
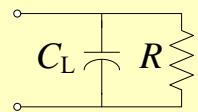


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### Capacitive Loads – Resistive Compensation

$$C_L = 1\text{pF}, R = 50\Omega$$



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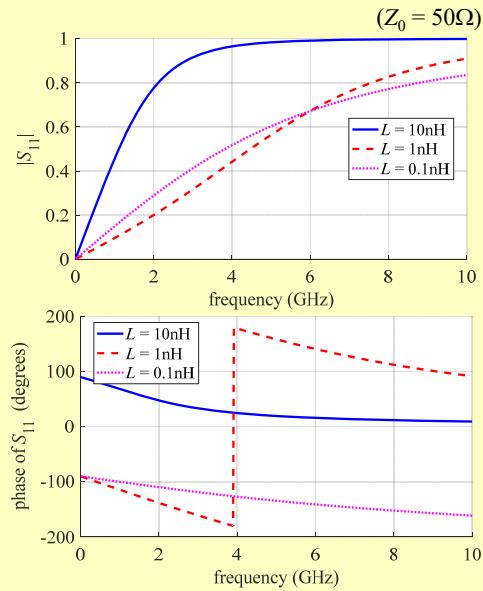
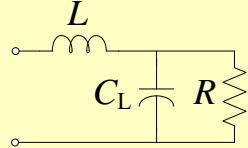
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### Capacitive Loads – Inductive Compensation

$$C_L = 1\text{pF}, R = 50\Omega$$



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### Matching Capacitive Loads – Example

Calculating  $R$  and  $L$ :

$$C_L = 1\text{pF}$$

$$f_{\max} = \frac{1}{4\pi Z_0 C_L} = 1.59 \text{ GHz}$$

Let  $f_0 = 1 \text{ GHz}$

$$B_L = 2\pi f_0 C_L = 6.283 \text{ mS} \quad 4Z_0^2 B_L^2 < 1$$

$$R = \frac{1 \pm \sqrt{1 - 4Z_0^2 B_L^2}}{2Z_0 B_L} = \begin{cases} 450.36 \Omega \\ 56.24 \Omega \end{cases}$$

$$X = \frac{B_L R^2}{1 + B_L^2 R^2} = \begin{cases} 141.48 \Omega \\ 17.67 \Omega \end{cases} \quad L = \frac{X}{2\pi f_0} = \begin{cases} 22.52 \text{ nH} \\ 2.81 \text{ nH} \end{cases}$$

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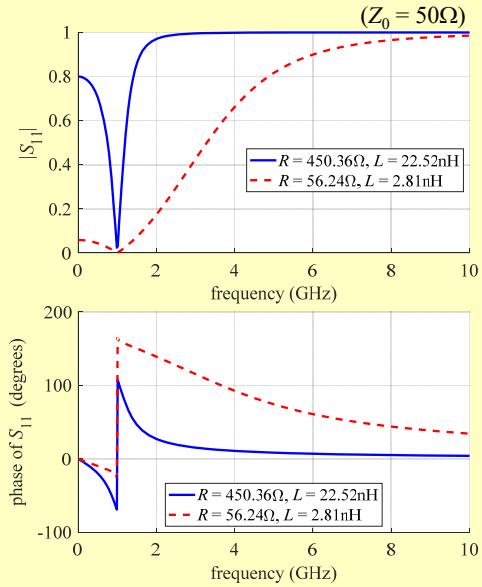
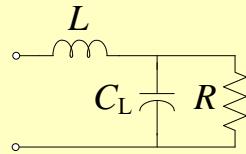
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### Matching Capacitive Loads – Example (cont.)

Using calculated  $R$  and  $L$ :

$$C_L = 1\text{pF}$$



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### Matching Capacitive Loads – Example (cont.)

Using an L-Section:

$$(Z_0 = 50\Omega)$$

Load ( $Z_L$ ):	L-Section Circuit 1:	L-Section Circuit 2:
 $C_L = 1\text{pF}$	 $Z_L$	 $Z_L$
$f_0 = 3\text{GHz}$	$B = \frac{X_L \pm \sqrt{R_L/Z_0} \sqrt{X_L^2 - R_L Z_0 + R_L^2}}{X_L^2 + R_L^2}$	$X = \pm \sqrt{R_L(Z_0 - R_L)} - X_L$
$Z_L = \frac{R}{1 + 2j\pi f_0 R C_L}$	$X = (Z_0 / R_L)[B(R_L^2 + X_L^2) - X_L]$	$B = \frac{X + X_L}{(X + X_L)^2 + R_L^2}$

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### Matching Capacitive Loads – Example (cont.)

Using an L-Section:

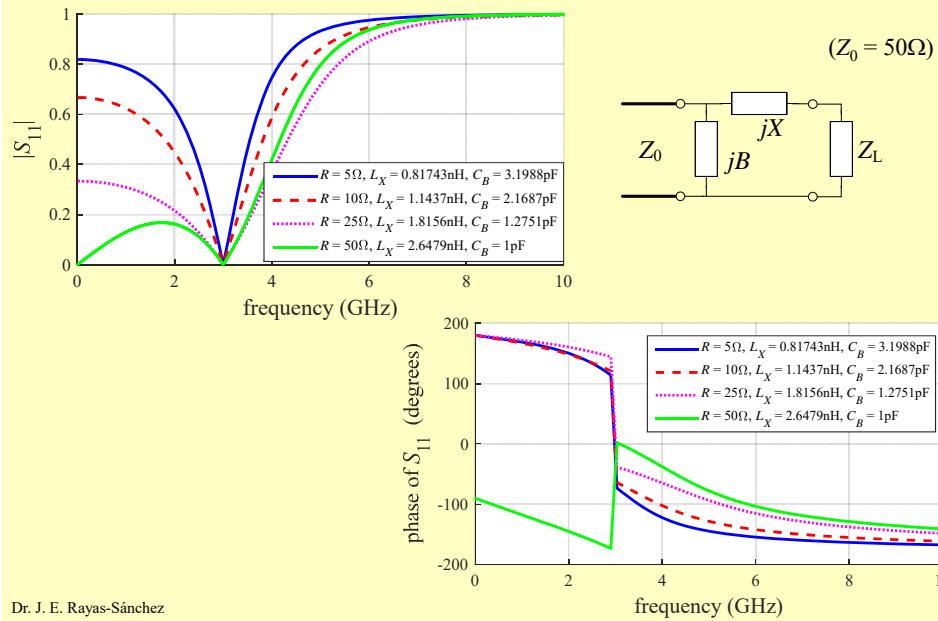
$(Z_0 = 50\Omega)$

$R (\Omega)$	$\text{Re}\{Z_L\} (\Omega)$	$\text{Im}\{Z_L\} (\Omega)$	Circuit	$X (\Omega)$	$B (\text{S})$	$L_X (\text{nH})$	$C_X (\text{pF})$	$L_B (\text{nH})$	$C_B (\text{pF})$
5	4.956	-0.46709	2	15.4082 -14.474	0.060295 -0.060295	0.81743	3.6653	0.87986	3.1988
10	9.6569	-1.8203	2	21.5583 -17.9177	0.040879 -0.040879	1.1437	2.9608	1.2978	2.1687
25	20.4572	-9.6402	2	34.224 -14.9436	0.024034 -0.024034	1.8156	3.5501	2.2073	1.2751
50	26.4793	-24.9562	2	49.9124 -3.5e-15	0.01885 -0.01885	2.6479	1.5e+16	2.8145	1
80	24.4353	-36.8475	2	61.8411 11.8539	0.020457 -0.020457	3.2808 0.62887		2.5933	1.0853
150	16.6771	-47.1533	2	70.7272 23.5794	0.028271 -0.028271	3.7522 1.2509		1.8765	1.4998
500	5.5663	-52.461	2	68.1878 36.7343	0.056507 -0.056507	3.6175 1.9488		0.93885	2.9978
1000	2.8066	-52.9028	2	64.4115 41.394	0.082013 -0.082013	3.4171 2.196		0.64687	4.3509

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### Matching Capacitive Loads – Example (cont.)



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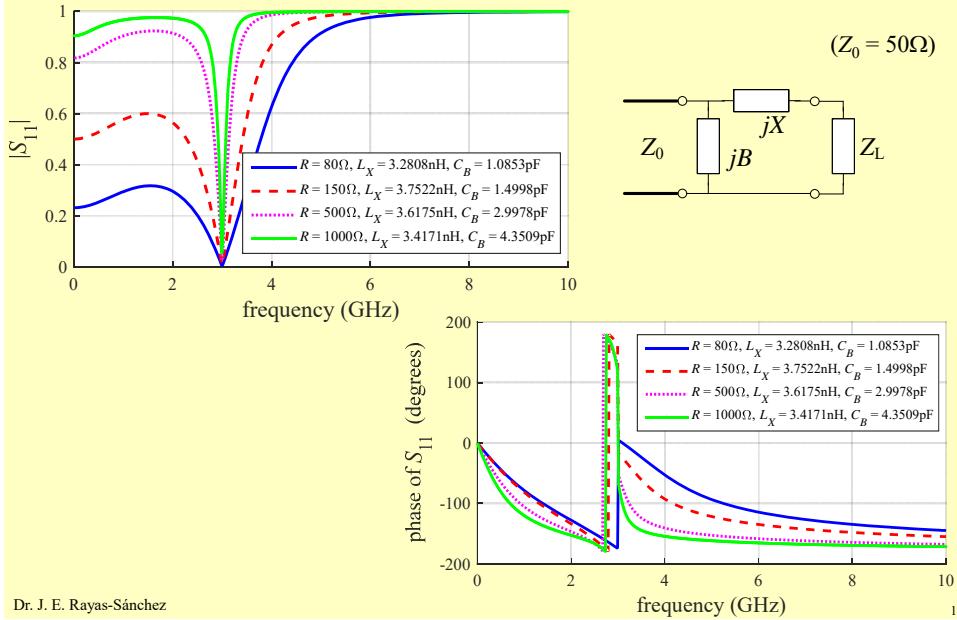
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### Matching Capacitive Loads – Example (cont.)



### Other Practical Techniques

H. Johnson, “High-speed digital design,” *IEEE Microwave Magazine*, vol. 12, pp. 42-50, Aug. 2011.



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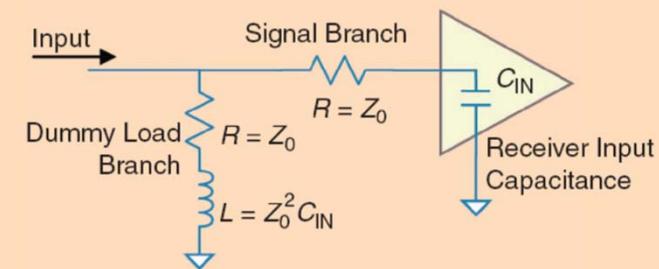
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### Other Practical Techniques (cont.)



**Figure 5.** The input impedance of this constant-resistance network, if perfectly implemented, equals  $Z_0$  at all frequencies.

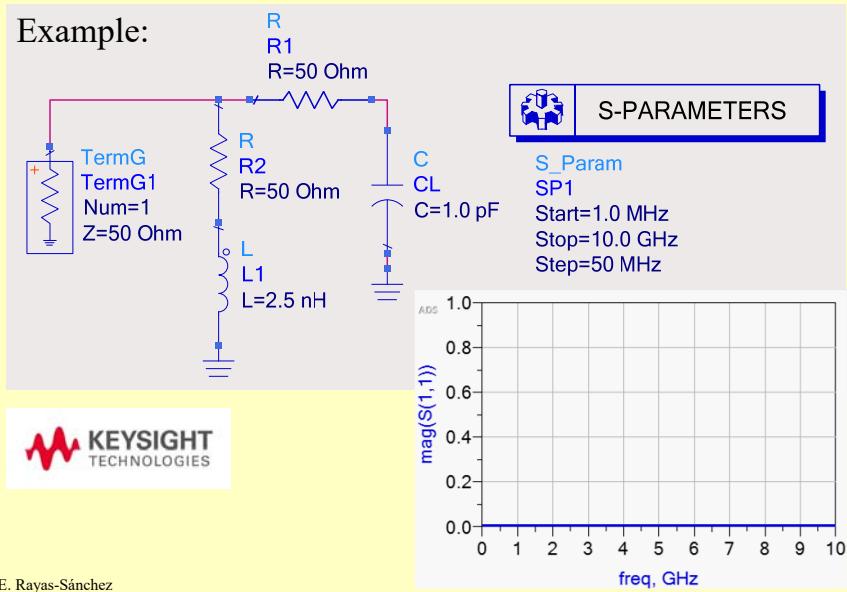
H. Johnson, "High-speed digital design," *IEEE Microwave Magazine*, vol. 12, pp. 42-50, Aug. 2011.

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### Other Practical Techniques (cont.)

Example:



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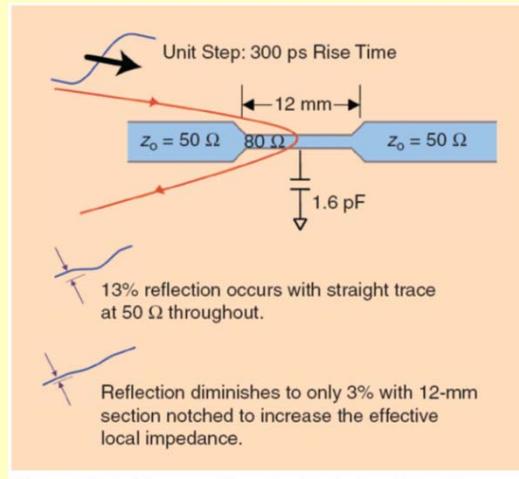
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### Other Practical Techniques (cont.)



**Figure 6.** A skinny section of printed circuit board trace mitigates the reflections from a small capacitive load.

H. Johnson, “High-speed digital design,” *IEEE Microwave Magazine*, vol. 12, pp. 42-50, Aug. 2011.

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