

BJT Fundamental Amplifier Configurations

Dr. José Ernesto Rayas Sánchez

1

Outline

- Small-signal middle-frequency equivalent circuits
- Common Emitter (CE)
- Common Base (CB)
- Common Collector (CC)
- Emitter Degeneration (ED)
- Comparison between CE, CB, CC and ED
- Distortion limits (dynamic range)

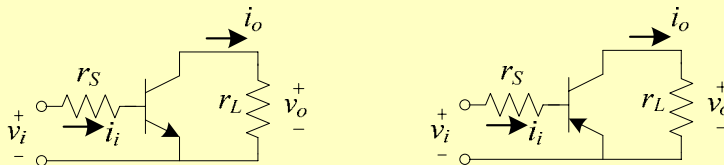
Small-Signal Mid-Frequency Equivalent Circuits

- To analyze an amplifier:
 1. Calculate the DC operating point (bias circuit)
 2. Obtain the small-signal equivalent circuit at medium frequencies
 3. Identify the amplifier's configuration (CE, CB, etc.)
 4. Analyze the small-signal equivalent circuit, either by using formulas or by deriving them

Dr. J. E. Rayas Sánchez

3

Common Emitter (CE)



$$A_v = \frac{-g_m(r_o \parallel r_L)r_\pi}{r_S + r_\pi} \quad R_i = r_S + r_\pi \quad R_o = r_o \quad A_i = \frac{-\beta r_o}{r_o + r_L}$$

$$\text{If } r_S \ll r_\pi, \quad A_v \approx -g_m(r_o \parallel r_L) \quad R_i \approx r_\pi$$

$$\text{If } r_L \ll r_o, \quad A_v \approx -g_m r_L \quad A_i \approx -\beta$$

Dr. J. E. Rayas Sánchez

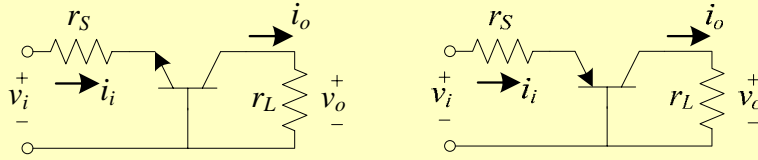
4

BJT Fundamental Amplifier Configurations

Dr. José Ernesto Rayas Sánchez

February 22, 2007

Common Base (CB)



$$A_v = \frac{r_L}{r_s + r_e \left(1 + \frac{r_L}{r_o}\right)} = \frac{g_m r_L r_o}{r_L + r_o (1 + g_m r_s)} \quad R_o = r_s + r_o (1 + g_m r_s)$$

$$A_i = \alpha \quad R_i = r_s + r_e \left(1 + \frac{r_L}{r_o}\right) = r_s + \frac{\alpha}{g_m} \left(1 + \frac{r_L}{r_o}\right)$$

If $r_L \ll r_o$,

$$A_v \approx \frac{r_L}{r_s + r_e} = \frac{g_m r_L}{1 + g_m r_s}$$

If $r_s \ll r_e$,

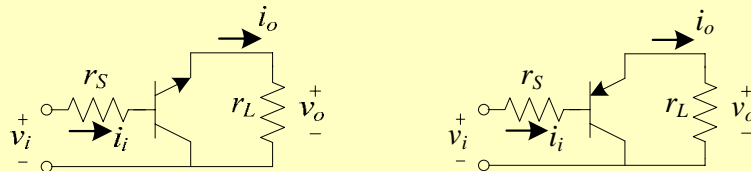
$$A_v \approx g_m (r_L \parallel r_o)$$

$$R_i = r_s + r_e = r_s + \frac{\alpha}{g_m}$$

Dr. J. E. Rayas Sánchez

5

Common Collector (CC)



$$A_v = \frac{(1 + \beta)(r_o \parallel r_L)}{r_s + r_\pi + (1 + \beta)(r_o \parallel r_L)} \quad A_v \approx \frac{\beta(r_o \parallel r_L)}{r_\pi + \beta(r_o \parallel r_L)} \quad A_i = \frac{(\beta + 1)r_o}{r_o + r_L}$$

$$R_i = r_s + r_\pi + (1 + \beta)(r_o \parallel r_L) \quad R_i \approx r_\pi + \beta(r_o \parallel r_L)$$

$$R_o = r_o \parallel (r_s + r_\pi) \parallel \left(\frac{1}{g_m} + \frac{r_s}{\beta}\right) \quad \text{If } r_o \gg r_s + r_\pi, \quad R_o \approx \frac{1}{g_m} + \frac{r_s}{\beta}$$

$$\text{If } r_L \ll r_o, \quad A_v \approx \frac{g_m r_L}{1 + g_m r_L} \quad R_i \approx r_\pi + \beta r_L$$

$$\text{If } \beta r_L \gg r_\pi, \quad A_v \approx 1$$

Dr. J. E. Rayas Sánchez

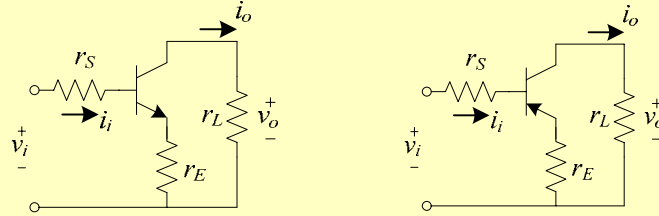
6

BJT Fundamental Amplifier Configurations

Dr. José Ernesto Rayas Sánchez

February 22, 2007

Emitter Degeneration (ED)



$$A_v = \frac{-g_m r_L}{1 + \frac{r_E}{r_\pi}(1 + \beta) + \frac{r_S}{r_\pi}}$$

If $r_S \ll r_E(1 + \beta)$,

$$A_v \approx \frac{-g_m r_L}{1 + g_m r_E}$$

If $g_m r_E \gg 1$,

Dr. J. E. Rayas Sánchez $A_v \approx -r_L / r_E$

$$R_i = r_S + r_\pi + (1 + \beta)r_E \quad A_i \approx -\beta$$

$$R_o = r_o + [r_E \parallel (r_S + r_\pi)] \left[1 + \frac{\beta r_o}{r_S + r_\pi} \right]$$

If $r_\pi \gg r_S$,

$$R_i \approx r_\pi + (1 + \beta)r_E$$

$$R_o \approx r_o [1 + g_m (r_E \parallel r_\pi)]$$

7

Simplified Formulas – Comparison

	CE	CB	CC	ED
A_v	$-g_m(r_o \parallel r_L)$	$\frac{g_m r_L}{1 + g_m r_S}$	$\frac{g_m r_L}{1 + g_m r_L}$	$\frac{-g_m r_L}{1 + g_m r_E}$
R_i	$r_S + r_\pi$	$r_S + \frac{1}{g_m} \left(1 + \frac{r_L}{r_o} \right)$	$r_\pi + \beta(r_o \parallel r_L)$	$r_\pi + \beta r_E$
R_o	r_o	$r_o(1 + g_m r_S)$	$\frac{1}{g_m} + \frac{r_S}{\beta}$	$r_o [1 + g_m (r_E \parallel r_\pi)]$
A_i	$-\beta$	α	β	$-\beta$

Dr. J. E. Rayas Sánchez

8

BJT Fundamental Amplifier Configurations

Dr. José Ernesto Rayas Sánchez

February 22, 2007

Over Simplified Formulas – Summary

	CE	CB	CC	ED
A_v	$-g_m r_L$	$g_m r_L$	1	$\frac{-r_L}{r_E}$
R_i	r_π	$r_S + \frac{1}{g_m}$	$r_\pi + \beta r_L$	$r_\pi + \beta r_E$
R_o	r_o	r_o	$\frac{1}{g_m}$	$r_o g_m (r_E \parallel r_\pi)$
A_i	$-\beta$	1	β	$-\beta$

Dr. J. E. Rayas Sánchez

9

Distortion Limits

- Distortion limits refer to the maximum output voltage swing for a linear operation (dynamic range)
- They depend on the amplifier configuration, the bias point, and the load line

- For CE amplifiers: $DL^+ = I_{CQ} r_L$ $DL^- = V_{CEQ} - V_{CE(sat)}$

- For CC amplifiers: $DL^+ = V_{CEQ} - V_{CE(sat)}$ $DL^- = I_{CQ} r_L$

- For CB amplifiers: $DL^+ = I_{CQ} r_L$ $DL^- = V_{CBQ} + V_{CBon}$

- For ED amplifiers: $DL^+ = I_{CQ} r_L$ $DL^- = (V_{CEQ} - V_{CE(sat)}) \frac{r_L}{r_L + r_E}$

Dr. J. E. Rayas Sánchez

10